

Cyber-physical systems under DoS attacks

A control systems perspective

Márcio J. Lacerda



1. Switched systems
2. Cyber-Physical systems
3. Control design for CPS under attacks
4. Extensions and future directions
5. Final remarks

Switched systems

Consider the following discrete-time switched system

$$x(k + 1) = A(\xi(k))x(k), \quad (1)$$

where $x \in \mathbb{R}^n$ is the state vector, and the switching rule is unknown *a priori*.

Consider the following discrete-time switched system

$$x(k+1) = A(\xi(k))x(k), \quad (1)$$

where $x \in \mathbb{R}^n$ is the state vector, and the switching rule is unknown *a priori*. The dynamic matrix can be written as

$$A(\xi(k)) = \sum_{i=1}^v \xi_i(k)A_i = \xi_1(k)A_1 + \xi_2(k)A_2 + \dots + \xi_v(k)A_v, \quad (2)$$

and the indicator function is defined as

$$\xi_i(k) = \begin{cases} 1, & \text{for } A_i \text{ (the } i\text{th mode is active)} \\ 0, & \text{otherwise} \end{cases} \quad \forall i = 1, \dots, v.$$

Consider the following discrete-time switched system

$$x(k+1) = A(\xi(k))x(k), \quad (1)$$

where $x \in \mathbb{R}^n$ is the state vector, and the switching rule is unknown *a priori*. The dynamic matrix can be written as

$$A(\xi(k)) = \sum_{i=1}^v \xi_i(k)A_i = \xi_1(k)A_1 + \xi_2(k)A_2 + \dots + \xi_v(k)A_v, \quad (2)$$

and the indicator function is defined as

$$\xi_i(k) = \begin{cases} 1, & \text{for } A_i \text{ (the } i\text{th mode is active)} \\ 0, & \text{otherwise} \end{cases} \quad \forall i = 1, \dots, v.$$

Stability

How can we certify that system (1) is globally asymptotically stable?

Theorem

The zero equilibrium of $x(k+1) = f_k(x(k))$ is globally uniformly asymptotically stable if there is a function $V : \mathbb{Z}^+ \times \mathbb{R}^n \rightarrow \mathbb{R}$ such that:

- ⊙ V is a positive-definite function, decreasing along the trajectories, and radially unbounded;
- ⊙ $\Delta V(k, x(k)) = V(k+1, x(k+1)) - V(k, x(k))$ is negative definite along the solutions of $x(k+1) = f_k(x(k))$.

One can say that the Lyapunov function is positive-definite, decreasing along the trajectories, and radially unbounded if $V(k, 0) = 0, \forall k \geq 0$ and

$$\beta_1 \|x(k)\|^2 \leq V(k, x(k)) \leq \beta_2 \|x(k)\|^2 \quad (3)$$

for all $x(k) \in \mathbb{R}^n$ and $k \geq 0$ with β_1 and β_2 positive scalars.

Switched Lyapunov function¹ $V(k, x(k)) = x(k)^T P(\xi(k))x(k)$.

Theorem

If there exist symmetric matrices P_1, \dots, P_v , such that

$$\begin{bmatrix} P_i & A_i^T P_j \\ P_j A_i & P_j \end{bmatrix} > 0, \quad \forall (i, j) \in \mathcal{J} \times \mathcal{J}, \quad (4)$$

where $\mathcal{J} = \{1, \dots, v\}$, then, the Lyapunov function $V(k, x(k)) = x(k)^T P(\xi(k))x(k)$ certify the stability of the switched system $x(k+1) = A(\xi(k))x(k)$.

¹J. Daafouz, P. Riedinger, and C. Iung, "Stability analysis and control synthesis for switched systems: a switched Lyapunov function approach," IEEE Transactions on Automatic Control, 2002.

Switched Lyapunov function¹ $V(k, x(k)) = x(k)^T P(\xi(k))x(k)$.

Theorem

If there exist symmetric matrices P_1, \dots, P_v , such that

$$\begin{bmatrix} P_i & A_i^T P_j \\ P_j A_i & P_j \end{bmatrix} > 0, \quad \forall (i, j) \in \mathcal{J} \times \mathcal{J}, \quad (4)$$

where $\mathcal{J} = \{1, \dots, v\}$, then, the Lyapunov function $V(k, x(k)) = x(k)^T P(\xi(k))x(k)$ certify the stability of the switched system $x(k+1) = A(\xi(k))x(k)$.

Idea of the proof:

$$\begin{aligned} \Delta(V) &= V(k+1, x(k+1)) - V(k, x(k)) < 0 \\ &= x(k+1)^T P(\xi(k+1))x(k+1) - x(k)^T P(\xi(k))x(k) < 0 \\ &= x(k)^T (A(\xi(k))P(\xi(k+1))A(\xi(k)) - P(\xi(k)))x(k) < 0 \end{aligned}$$

¹J. Daafouz, P. Riedinger, and C. Iung, "Stability analysis and control synthesis for switched systems: a switched Lyapunov function approach," IEEE Transactions on Automatic Control, 2002.

Structured Lyapunov functions

By employing an augmented state vector in the Lyapunov function²

$$V(k) = \begin{bmatrix} x(k) \\ x(k+1) \\ \vdots \\ x(k+N-1) \end{bmatrix}^T \Psi \begin{bmatrix} x(k) \\ x(k+1) \\ \vdots \\ x(k+N-1) \end{bmatrix}$$

with

$$\Psi = \text{blkdiag}(P_1(\xi(k)), P_2(\xi(k+1)), \dots, P_N(\xi(k+N-1))),$$

we are able to derive necessary and sufficient conditions to certify the stability of the switched system $x(k+1) = A(\xi(k))x(k)$.

²M. J. Lacerda and T. D. S. Gomide. "Stability and stabilisability of switched discrete-time systems based on structured Lyapunov functions". IET Control Theory & Applications, 2020.

Structured Lyapunov functions

The use of Lyapunov functions with non-monotonic terms³

$V_i(x(k)) = x(k)^T P_i(\xi(k))x(k)$ can also lead to necessary and sufficient conditions⁴ to certify the stability of the switched system $x(k+1) = A(\xi(k))x(k)$.

$$\sum_{i=j}^N V_i(x(k)) > 0, \quad j = 1, \dots, N,$$

$$V_1(x(k+1)) - V_1(x(k)) + V_2(x(k+2)) - V_2(x(k)) + \dots + V_N(x(k+N)) - V_N(x(k)) < 0.$$

³A. A. Ahmadi and P. A. Parrilo. "Non-monotonic Lyapunov functions for stability of discrete time nonlinear and switched systems." 47th IEEE Conference on Decision and Control, 2008.

⁴M. J. Lacerda and T. D. S. Gomide. "Stability and stabilisability of switched discrete-time systems based on structured Lyapunov functions". IET Control Theory & Applications, 2020.

Non-monotonic terms

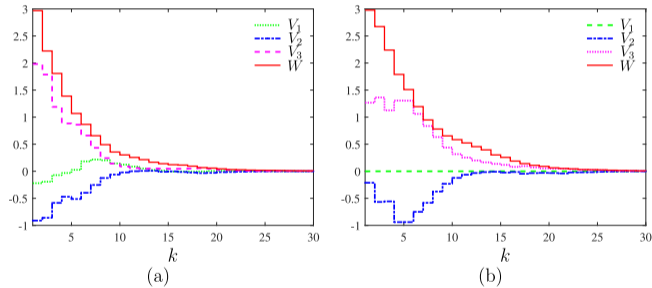


Fig. 1. Evolution of functions $V_1(x_k)$ (dotted green line), $V_2(x_k)$ (dashed dotted blue line), $V_3(x_k)$ (magenta dashed line) and the Lyapunov function $W(x_k)$ (straight red line) - Example 2. (For interpretation of the references to color in this figure legend, the reader is referred to the web version of this article.)

Consider the following discrete-time switched system

$$x(k+1) = A(\xi(k))x(k) + B(\xi(k))u(k), \quad (5)$$

where $x \in \mathbb{R}^n$ is the state vector, and $u \in \mathbb{R}^{n_u}$ is the control input. The switching rule is unknown *a priori*, but it is considered to be available in real-time.

Consider the following discrete-time switched system

$$x(k+1) = A(\xi(k))x(k) + B(\xi(k))u(k), \quad (5)$$

where $x \in \mathbb{R}^n$ is the state vector, and $u \in \mathbb{R}^{n_u}$ is the control input. The switching rule is unknown *a priori*, but it is considered to be available in real-time.

State-feedback control

Design a switching state-feedback control law

$$u(k) = K(\xi(k))x(k),$$

where $K(\xi(k)) \in \mathbb{R}^{n_u \times n}$ stabilizes the closed-loop system

$$x(k+1) = (A(\xi(k)) + B(\xi(k))K(\xi(k)))x(k). \quad (6)$$

Theorem

If there exist symmetric matrices $P_j \in \mathbb{R}^{n \times n}$, $j = 1, \dots, N$, and matrices $X(\xi(k)) \in \mathbb{R}^{n \times n}$ and $Z(\xi(k)) \in \mathbb{R}^{n_u \times n}$, also defined as in (2), such that the following inequalities are satisfied

$$\sum_{m=j}^N P_m > 0, \quad j = 2, \dots, N$$

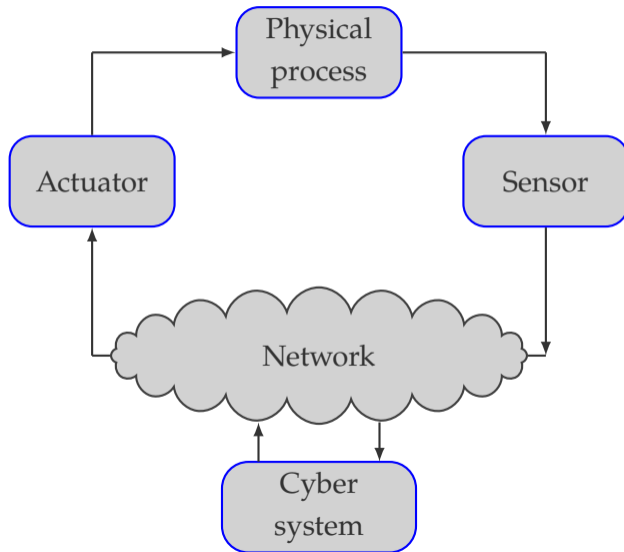
$$\begin{bmatrix} -P_1 & A_{i_1} X_{i_1} + B_{i_1} Z_{i_1} & \cdots & 0 & 0 \\ \star & P_1 - P_2 - X_{i_1} - X_{i_1}^T & \ddots & \vdots & \vdots \\ 0 & 0 & \ddots & A_{i_{N-1}} X_{i_{N-1}} + B_{i_{N-1}} Z_{i_{N-1}} & 0 \\ \vdots & \vdots & \ddots & P_{N-1} - P_N - X_{i_{N-1}} - X_{i_{N-1}}^T & A_{i_N} X_{i_N} + B_{i_N} Z_{i_N} \\ 0 & 0 & \cdots & \star & P_N - X_{i_N} - X_{i_N}^T \end{bmatrix} < 0$$

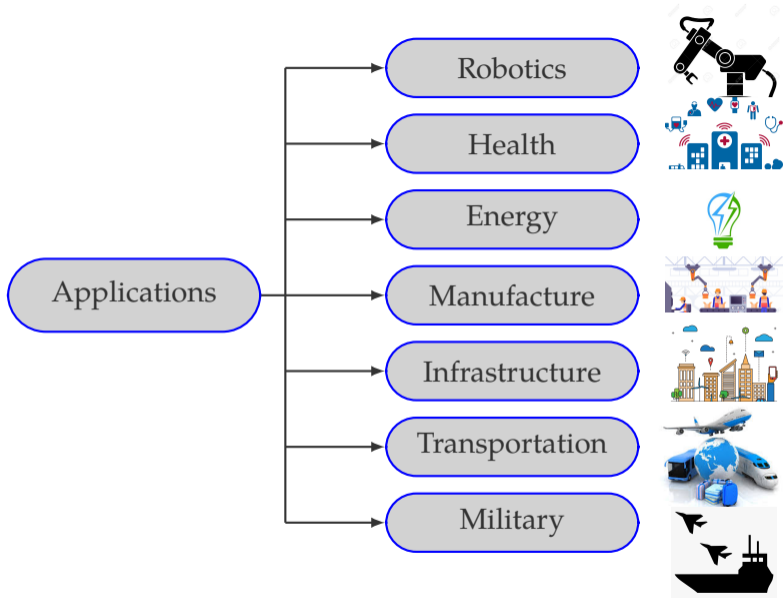
$$\forall (i_1, i_2, \dots, i_N) \in \underbrace{\mathcal{F} \times \mathcal{F} \dots \mathcal{F}}_{N \text{ times}}$$

then, $K_{i_m} = Z_{i_m} X_{i_m}^{-1}$ are the state feedback control gains assuring that the closed loop system $x(k+1) = (A(\xi(k)) + B(\xi(k))K(\xi(k)))x(k)$ is asymptotically stable.

Cyber-Physical systems

What is a cyber-physical system (CPS)?



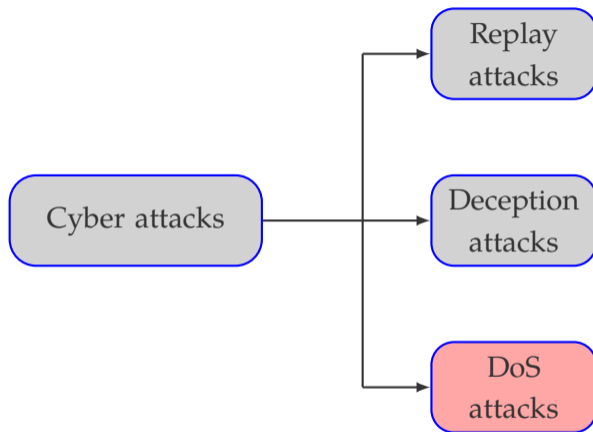




Source: Data & Analytics Facility for National Infrastructure (DAFNI) to advance UK infrastructure research.

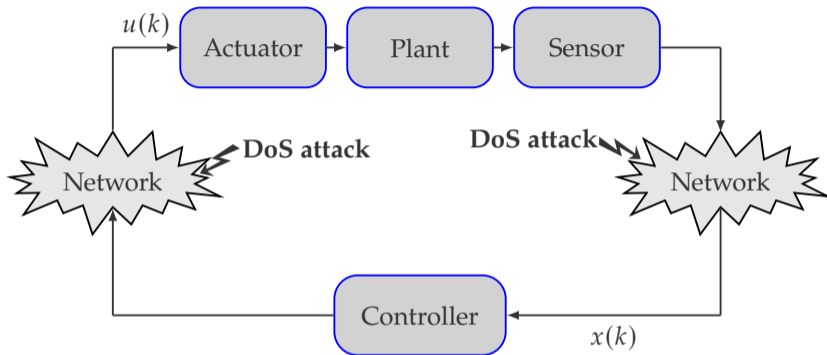
Virtual organization of a CPS



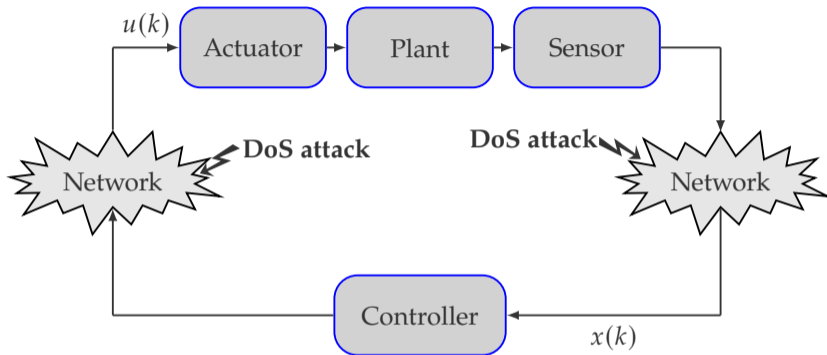


⁵D. Ding, Q. L. Han, Y. Xiang, X. Ge, and X.-M. Zhang, "A survey on security control and attack detection for industrial cyber-physical systems," *Neurocomputing*, vol. 275, pp. 1674 – 1683, 2018.

Structure of a CPS under DoS attacks.



Structure of a CPS under DoS attacks.



Problem

Does the designed controller ensure the stability of the closed-loop system under the presence of DoS attacks?

Control design for CPS under attacks

$$x(k + 1) = A(\alpha)x(k) + B(\alpha)u(k) \quad (7)$$

where $x \in \mathbb{R}^n$ is the state vector, and $u \in \mathbb{R}^{n_u}$ the control input.

$$x(k+1) = A(\alpha)x(k) + B(\alpha)u(k) \quad (7)$$

where $x \in \mathbb{R}^n$ is the state vector, and $u \in \mathbb{R}^{n_u}$ the control input.

Scenario

- ⊙ The matrices $A(\alpha)$ and $B(\alpha)$ belong to an **uncertain domain**.

$$\begin{bmatrix} A(\alpha) & B(\alpha) \end{bmatrix} = \sum_{i=1}^V \alpha_i \begin{bmatrix} A_i & B_i \end{bmatrix}, \quad \alpha \in \Lambda,$$

- ⊙ V denotes the number of vertices of the polytope and Λ is the unit simplex

$$\Lambda = \left\{ \alpha \in \mathbb{R}^V : \sum_{i=1}^V \alpha_i = 1, \alpha_i \geq 0 \right\}.$$

Consider a discrete-time uncertain system with matrices

$$A = \begin{bmatrix} 1 & 0.1 \\ 0 & 1 - 0.1\delta \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0.1\kappa \end{bmatrix},$$

where $0.1s^{-1} \leq \delta \leq 10s^{-1}$, and $\kappa = 0.787rad^{-1}V^{-1}s^{-2}$.

Disregarding the existence of attack the following state-feedback control gain stabilizes the system

$$K = \begin{bmatrix} -6.6145 & -7.4944 \end{bmatrix}.$$

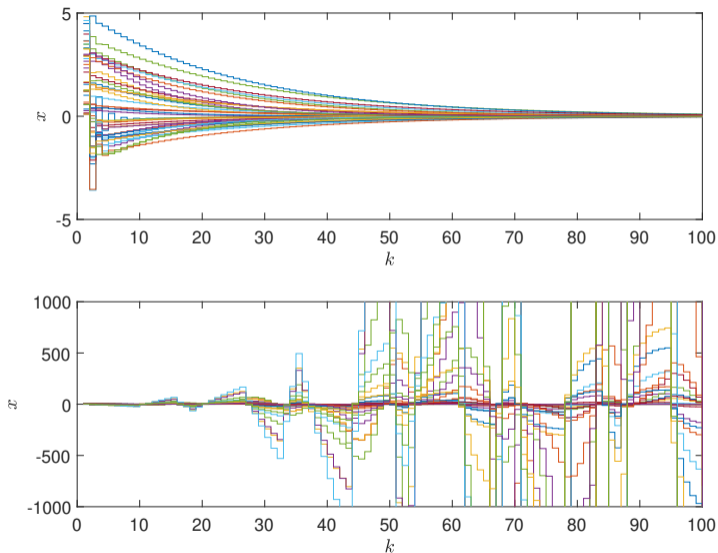


Figure 1: Trajectories for the closed-loop states in the absence of attacks (upper), and during the presence of DoS attack (lower).

How can we design a control strategy capable of ensuring the stability of the closed-loop uncertain system under the presence of DoS attacks?

- ⊙ We need to construct a model that takes into account the presence of DoS attacks. Different control strategies can be employed⁶:
 - Hold strategy
 - Zero strategy
 - Packet of different controllers
- ⊙ By using the Lyapunov theory, the design conditions will be written in the form of LMIs.
- ⊙ The designed controllers will be capable of ensuring the stability of the closed-loop uncertain system under the presence of DoS attacks.

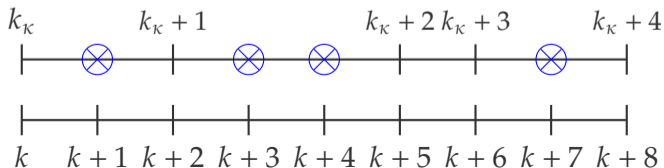
⁶L. Schenato, "To Zero or to Hold Control Inputs With Lossy Links?," IEEE Transactions on Automatic Control, 2009.

Problem Formulation: DoS modelling

Assumption: The duration of the DoS attack is bounded by the maximum number of consecutive control inputs samples that do not get to the actuator, being this number denoted by N .

- ⊙ Switching signal $\sigma(k_\kappa)$ that assume values in $M \triangleq \{0, 1, \dots, N\}$
- ⊙ A new time scale k_κ that represents the time instant when the updated control input reaches the actuator

$$k_\kappa + 1 = k_\kappa + \sigma(k_\kappa) + 1, \quad k_0 = 0, \quad \sigma(k_\kappa) = \{1, 2, 0, 1\}$$



Problem Formulation: Hold Strategy

- ⊙ The same control input $u(k) = Kx(k)$ available to the actuator is successively applied until the end of the attack (next successful transmission).

$N = 1$

$$x(k+1) = A(\alpha)x(k) + B(\alpha)Kx(k),$$

$$x(k+2) = A(\alpha)x(k+1) + B(\alpha)Kx(k),$$

$$\rightarrow x(k+2) = A(\alpha)^2x(k) + A(\alpha)B(\alpha)Kx(k) + B(\alpha)Kx(k),$$

Problem Formulation: Hold Strategy

- ⊙ The same control input $u(k) = Kx(k)$ available to the actuator is successively applied until the end of the attack (next successful transmission).

$N = 1$

$$x(k+1) = A(\alpha)x(k) + B(\alpha)Kx(k),$$

$$x(k+2) = A(\alpha)x(k+1) + B(\alpha)Kx(k),$$

$$\rightarrow x(k+2) = A(\alpha)^2x(k) + A(\alpha)B(\alpha)Kx(k) + B(\alpha)Kx(k),$$

$N = 2$

$$x(k+3) = A(\alpha)x(k+2) + B(\alpha)Kx(k),$$

$$\begin{aligned} \rightarrow x(k+3) &= A(\alpha)^3x(k) + A(\alpha)^2B(\alpha)Kx(k) + A(\alpha)B(\alpha)Kx(k) \\ &\quad + B(\alpha)Kx(k). \end{aligned}$$

Problem Formulation: Zero Strategy

- ⊙ The control input is set to zero until the end of the attack (next successful transmission).

$$N = 1$$

$$\begin{aligned}x(k+1) &= A(\alpha)x(k) + B(\alpha)Kx(k), \\x(k+2) &= A(\alpha)x(k+1) \\ \rightarrow x(k+2) &= A(\alpha)^2x(k) + A(\alpha)B(\alpha)Kx(k),\end{aligned}\tag{8}$$

Problem Formulation: Zero Strategy

- ⊙ The control input is set to zero until the end of the attack (next successful transmission).

$$N = 1$$

$$\begin{aligned}x(k+1) &= A(\alpha)x(k) + B(\alpha)Kx(k), \\x(k+2) &= A(\alpha)x(k+1) \\ \rightarrow x(k+2) &= A(\alpha)^2x(k) + A(\alpha)B(\alpha)Kx(k),\end{aligned}\tag{8}$$

$$N = 2$$

$$\begin{aligned}x(k+3) &= A(\alpha)x(k+2), \\ \rightarrow x(k+3) &= A(\alpha)^3x(k) + A(\alpha)^2B(\alpha)Kx(k).\end{aligned}$$

Problem Formulation: Packet Strategy

- ⊙ Different control inputs $u(k+i) = K_i x(k)$ are available to the actuator before an attack starts in $k+1$. These inputs are successively applied until the end of the attack (next successful transmission).

$N = 1$

$$x(k+1) = A(\alpha)x(k) + B(\alpha)K_0x(k),$$

$$x(k+2) = A(\alpha)x(k+1) + B(\alpha)K_1x(k),$$

$$\rightarrow x(k+2) = A(\alpha)^2x(k) + A(\alpha)B(\alpha)K_0x(k) + B(\alpha)K_1x(k),$$

Problem Formulation: Packet Strategy

- ⊙ Different control inputs $u(k+i) = K_i x(k)$ are available to the actuator before an attack starts in $k+1$. These inputs are successively applied until the end of the attack (next successful transmission).

$N = 1$

$$x(k+1) = A(\alpha)x(k) + B(\alpha)K_0x(k),$$

$$x(k+2) = A(\alpha)x(k+1) + B(\alpha)K_1x(k),$$

$$\rightarrow x(k+2) = A(\alpha)^2x(k) + A(\alpha)B(\alpha)K_0x(k) + B(\alpha)K_1x(k),$$

$N = 2$

$$x(k+3) = A(\alpha)x(k+2) + B(\alpha)K_2x(k),$$

$$\begin{aligned} \rightarrow x(k+3) &= A(\alpha)^3x(k) + A(\alpha)^2B(\alpha)K_0x(k) + A(\alpha)B(\alpha)K_1x(k) \\ &\quad + B(\alpha)K_2x(k). \end{aligned}$$

Packet of controllers

$$U(k) = \begin{bmatrix} u(k) \\ u(k+1) \\ \vdots \\ u(k+N) \end{bmatrix} = \begin{bmatrix} K_0 x(k) \\ K_1 x(k) \\ \vdots \\ K_N x(k) \end{bmatrix}, \quad (9)$$

is the package that gets to the actuator side every time that the communications channels are free of the attack.

⁷P. S. P. Pessim and M. J. Lacerda, "State-Feedback Control for Cyber-Physical LPV Systems Under DoS Attacks." *IEEE Control Systems Letters*, 2021.

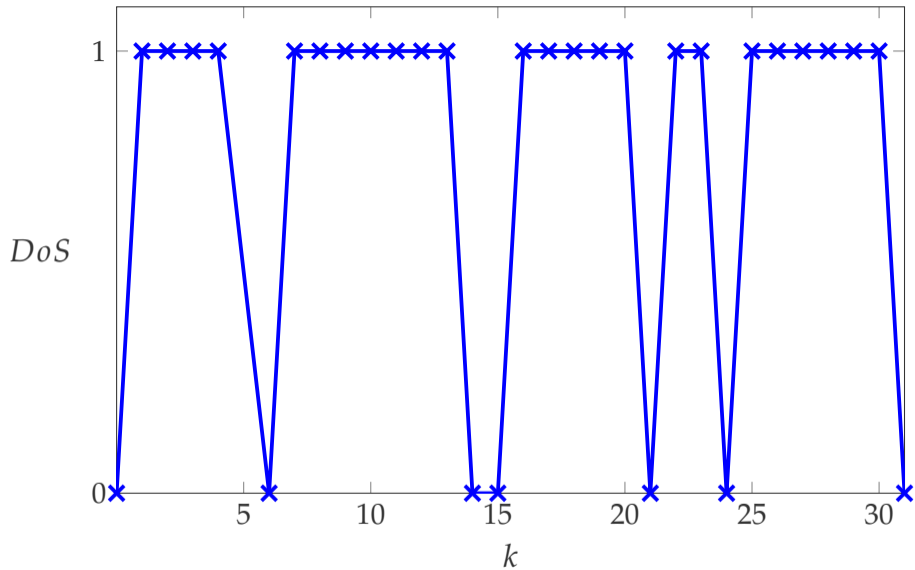


Figure 2: "1" - presence of DoS attacks and "0" - absence of DoS attacks. Sequence of attacks $\sigma(k_\kappa) = \{4, 7, 0, 5, 2, 6, \dots\}$

$$U_1(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix}, U_2(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix}, U_3(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix},$$

$$U_4(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix}, U_5(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix}, U_6(k_\kappa) = \begin{bmatrix} u(k_\kappa) \\ u(k_\kappa + 1) \\ u(k_\kappa + 2) \\ u(k_\kappa + 3) \\ u(k_\kappa + 4) \\ u(k_\kappa + 5) \\ u(k_\kappa + 6) \\ u(k_\kappa + 7) \end{bmatrix}.$$

Problem Formulation: Switched System

- ⊙ Case 0: DoS-free case

$$x(k_{\kappa} + 1) = (A(\alpha) + B(\alpha)K_0) x(k_{\kappa}),$$

$$x(k_{\kappa} + 1) = F_0(\alpha)x(k_{\kappa})$$

- ⊙ Case 1: The DoS attack occurs during one time-instant

$$x(k_{\kappa} + 1) = (A(\alpha)^2 + A(\alpha)B(\alpha)K_0 + B(\alpha)K_1) x(k_{\kappa}),$$

$$x(k_{\kappa} + 1) = F_1(\alpha)x(k_{\kappa}) = (A(\alpha)F_0(\alpha) + B(\alpha)K_1) x(k_{\kappa}).$$

- ⊙ Case 2: The DoS attack occurs during two time-instants

$$x(k_{\kappa} + 1) = F_2(\alpha)x(k_{\kappa}) = (A(\alpha)F_1(\alpha) + B(\alpha)K_2) x(k_{\kappa}).$$

Problem Formulation: Switched System

A generic formulation is given as follows

$$F_i(\alpha) = A(\alpha)F_{i-1}(\alpha) + B(\alpha)K_i,$$

$i = 1, \dots, N$, with $F_0(\alpha) = A(\alpha) + B(\alpha)K_0$. These matrices are used to construct the following switched system with $N + 1$ modes.

$$x(k_{\kappa} + 1) = F_{\sigma(k_{\kappa})}x(k_{\kappa}).$$

Considering the indicator function $\xi(k_{\kappa}) = [\xi_0(k_{\kappa}), \dots, \xi_N(k_{\kappa})]^T$

$$x(k_{\kappa} + 1) = F(\xi(k_{\kappa}))x(k_{\kappa}), \quad \xi_i(k_{\kappa}) = \begin{cases} 1, & \text{if } \sigma(k_{\kappa}) = i \\ 0, & \text{otherwise} \end{cases}$$

with $F(\xi(k_{\kappa})) = \xi_0(k_{\kappa})F_0 + \xi_1(k_{\kappa})F_1 + \dots + \xi_N(k_{\kappa})F_N$.

How to design the gain matrices?

Existence of a Lyapunov function $V(x_{k_k})$, that is positive definite, and has its time rate of change negative definite along the trajectories, i.e., $\Delta V(x_{k_k}) < 0$.

How to design the gain matrices?

Existence of a Lyapunov function $V(x_{k_k})$, that is positive definite, and has its time rate of change negative definite along the trajectories, i.e., $\Delta V(x_{k_k}) < 0$.

Moreover, we need to employ

1. Change of variables.
2. Congruence transformation.
3. Schur complement.
4. Linear Matrix Inequalities.

Theorem

If there exist symmetric positive definite matrices $Q_i \in \mathbb{R}^{n \times n}$, matrices $X \in \mathbb{R}^{n \times n}$ and $Z_i \in \mathbb{R}^{n_u \times n}$, such that

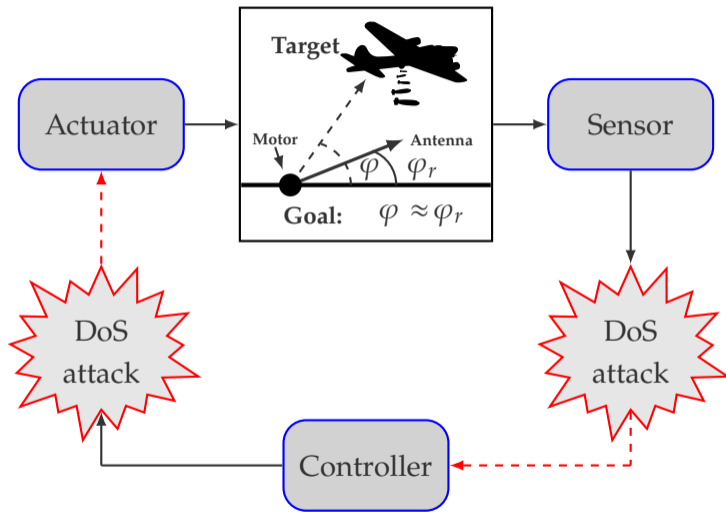
$$\begin{bmatrix} -Q_i(\alpha) & \star \\ \Psi_i & Q_j(\alpha) - X - X^T \end{bmatrix} < 0, \quad (10)$$

where

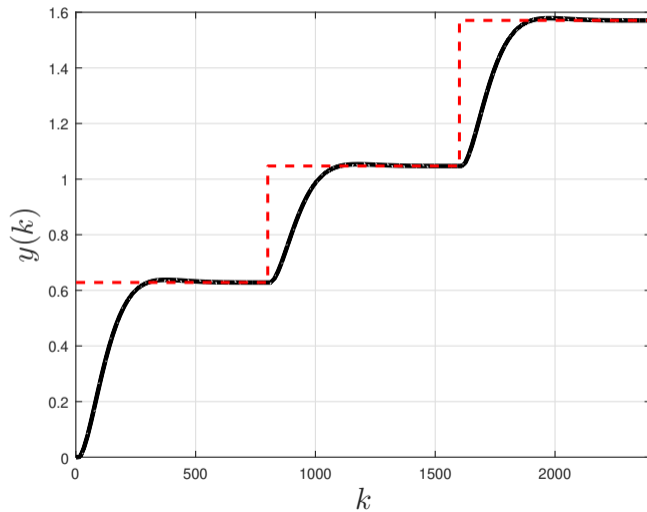
$$\Psi_i = A(\alpha)^{i+1}X + \sum_{m=0}^i A(\alpha)^m B(\alpha)Z_{i-m}, \quad (11)$$

with $A(\alpha)^0 = I_n$, $i, j \in M$, $M \triangleq \{0, 1, \dots, N\}$, then $K_i = Z_i X^{-1}$ are the state-feedback control gains that assure the closed-loop system (7) is asymptotically stable.

Example: angular positioning system



Example: angular positioning system $N = 14$



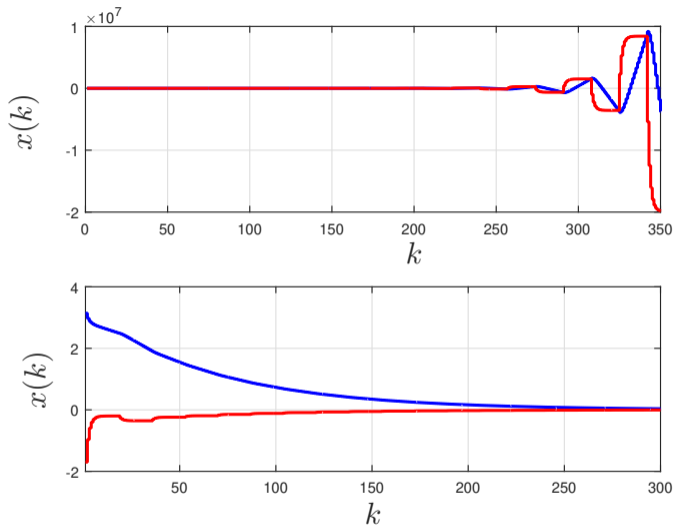


Figure 3: Trajectories of the system considering the presence of attack $N = 16$ in the design conditions (bottom) and disregarding the presence of attack in control design (top).

Problems addressed

- ⊙ Output-feedback control for LPV systems⁸.
- ⊙ \mathcal{H}_∞ performance for LPV systems⁹.
- ⊙ \mathcal{H}_2 performance for uncertain systems¹⁰.

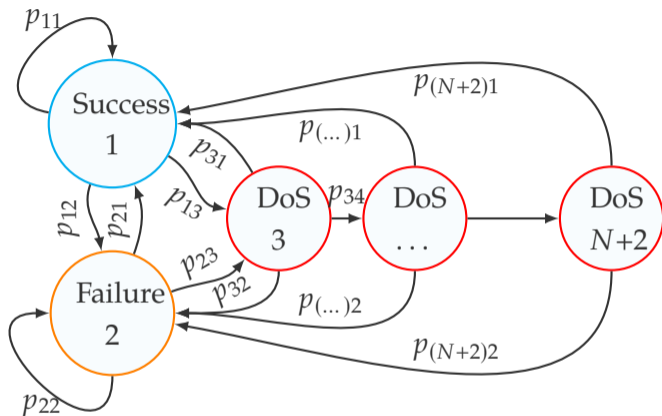
⁸P. S. P. Pessim and M. L. C. Peixoto and R. M. Palhares and M. J. Lacerda, "Static output-feedback control for Cyber-physical LPV systems under DoS attacks." *Information Sciences*, 2021.

⁹P. S. P. Pessim and M. J. Lacerda, "On the robustness of Cyber-physical LPV systems under DoS attacks." *Journal of the Franklin Institute*, 2022.

¹⁰P. M. Oliveira and J. M. Palma and M. J. Lacerda. " \mathcal{H}_2 state-feedback control for discrete-time cyber-physical uncertain systems under DoS attacks," *Applied Mathematics and Computation*, 2022.

Extensions and future directions

A model that includes **DoS attack**+**packet loss** for control design¹¹.



¹¹P. M. Oliveira and J. M. Palma and M. J. Lacerda. "Control Design for an Unreliable Markovian Network Susceptible to Denial-of-Service Attacks", IEEE Transactions on Circuits and Systems II: Express Briefs, 2024.

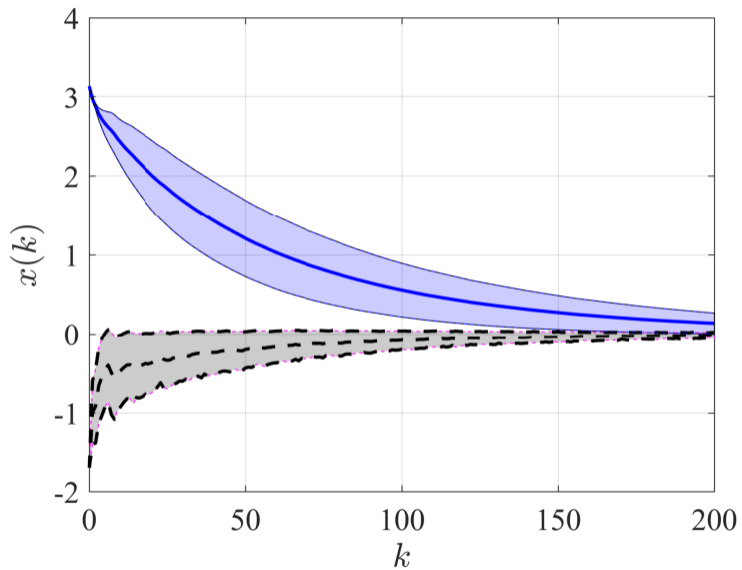
Consider the same example with

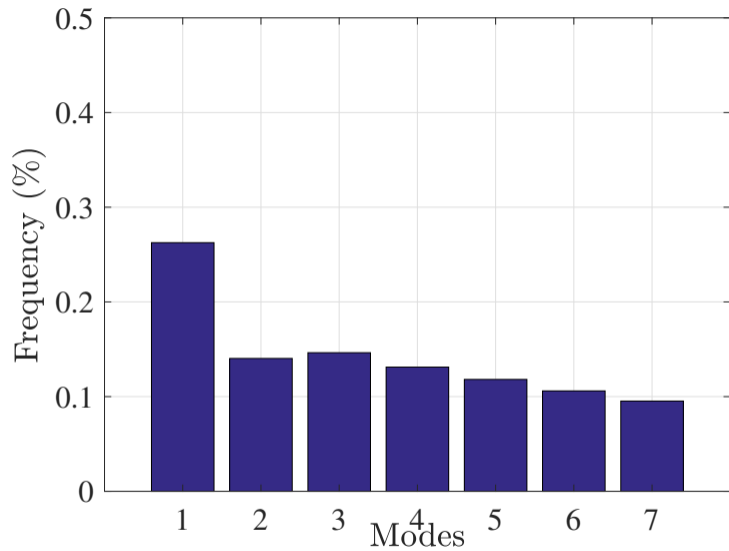
$$A(\alpha) = \begin{bmatrix} 1 & 0.1 \\ 0 & 1 - 0.1\delta \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0.1\kappa \end{bmatrix},$$

where $0.1s^{-1} \leq \delta \leq 10s^{-1}$, and $\kappa = 0.787rad^{-1}V^{-1}s^{-2}$. In this approach we need to take into account the transition probability matrix.

$$\Psi = \begin{bmatrix} 0.5 & c & d & 0 & 0 & 0 & 0 \\ 0.4 & ? & ? & 0 & 0 & 0 & 0 \\ 0.05 & 0.05 & 0 & 0.9 & 0 & 0 & 0 \\ 0.05 & 0.05 & 0 & 0 & 0.9 & 0 & 0 \\ 0.05 & 0.05 & 0 & 0 & 0 & 0.9 & 0 \\ 0.05 & 0.05 & 0 & 0 & 0 & 0 & 0.9 \\ 0.5 & 0.5 & 0 & 0 & 0 & 0 & 0 \end{bmatrix}, \quad (12)$$

where $c = \begin{bmatrix} 0.05 & 0.15 \end{bmatrix}$ and $d = \begin{bmatrix} 0.35 & 0.45 \end{bmatrix}$.





Secure control

- ⊙ Safety for cyber-physical systems under attacks using control barrier function.
- ⊙ Filter design for attack detection.
- ⊙ Constrained control input such as saturation.
- ⊙ Replay attacks and false data injection attacks.
- ⊙ Hybrid model for the CPS under attack.

Final remarks



CONTROL FOR
SOCIETAL-SCALE
CHALLENGES:
ROAD MAP
2030

Anuradha M. Annaswamy
Karl H. Johansson | George J. Pappas

Emerging Methodologies

- ⊙ Safety Critical systems
- ⊙ Resilient cyber-physical systems
- ⊙ Cyber-physical human systems

- ⊙ CPS present opportunities and new challenges for control design.
- ⊙ Control theory can contribute to safety in CPS.

- ⊙ CPS present opportunities and new challenges for control design.
- ⊙ Control theory can contribute to safety in CPS.

Thank you!

m.lacerda@londonmet.ac.uk